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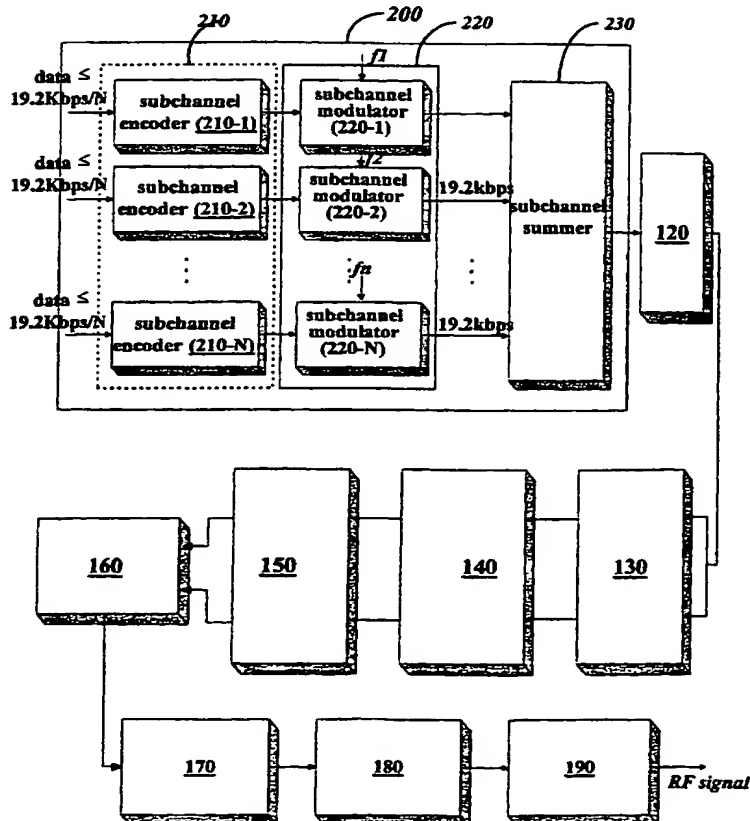
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(54) Title: APPARATUS AND METHOD FOR EXPANDING CHANNELS IN CDMA SYSTEM



(57) Abstract: This present invention relates to an apparatus and method for increasing channel, using multiple modulation of subchannels of low data rates, in a code division multiple access (CDMA) transmit modulator. In order to obtain a plurality of subchannels of low data rates within a traffic channel, a plurality of subchannel encoders are used such that each subchannel data is encoded independently and is then channelized using a plurality of Walsh orthogonal codes. The plurality of channelized subchannel data are combined in a signal and are then transmitted through a traffic channel.



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DESCRIPTION

APPARATUS AND METHOD FOR EXPANDING CHANNELS IN CDMA SYSTEM

1. Technical Field

5 The present invention relates to channels of code division multiple access systems and, more particularly, to an apparatus and method for increasing available channels in a code division multiple access (CDMA) transmit modulator.

2. Background Art

10 Fig. 1 shows a block diagram of a code division multiple access (CDMA) transmit modulator of the conventional art. The transmit modulator comprises a channel encoder 110, which includes a convolution encoder 111, a symbol repeater 112 and an interleaver 113, for convolutionally encoding with
15 repetition and interleaving the input data from a mobile station; a channel modulator 120, which includes a Walsh code combiner 121 and a Walsh code generator 122, for combining the output of channel encoder 100 with a Walsh code such that this channel is distinguishable from the other traffic
20 channels in the allocated frequency assignment; a pseudo noise (PN) code combining unit 130, including first and second PN code combiners 131 and 132, for combining the output of Walsh code combiner 121 and both of in-phase (I) and
quadrature (Q) channel PN sequences, PN_I and PN_Q, which are
25 generated and determined by a predetermined PN offset value such that multiple cell-sites or sectors using the same frequency assignment are distinguished from one another; a lowpass filter 140, including first and second digital finite impulse response (FIR) filters 141 and 142, for filtering out

high frequency components of the outputs of the first and second PN code combiners 131 and 132 and flattens them in the frequency assignment; an analog signal modulator 150, including first and second mixers 151 and 152, for multiplying the D/A-converted signals of the first and second FIR filters 141 and 142 with sine and cosine functions and producing the modulated analog signals; an analog signal summer 160 for summing the outputs of analog signal modulator 150; an intermediate frequency (IF) modulator 170 for modulating the signal of analog signal summer 160 by using the quadrature phase shift key (QPSK) modulation; a frequency up-converter 180 for up-converting the IF signal into a radio frequency (RF) signal; and an RF transmitter 190 for amplifying and radiating the RF signal through an antenna system.

The operation of the transmit modulator of the conventional art, shown in FIG. 1, is described below in detail.

The channel input data from a vocoder is convolutionally encoded for error correction by convolutional encoder 111. The encoding rate of convolutional encoder 111 is twice the input data rate. For example, for the input data symbols whose rate is 9,600 bits per second (bps), convolutional encoder 111 outputs data symbols whose rate is 19,200 bps. The encoded data is then provided as input to symbol repeater 112.

Depending on the input data rate, the symbols are repeated by symbol repeater 112 in order that the rate of the resulting output of symbol repeater 112 becomes 19,200 bps. That is, symbols are repeated for the input data of low rates. For example, if the input data rate is 9,600 bps, the symbols are repeated twice, if the input data rate is 4,800 bps, they are repeated four times, and so on. The repeated symbols are inputted to interleaver 113. Interleaving is done by interleaver 113 with reference to a predetermined sequence to spread possible burst errors into random errors.

Walsh code generator 122 generates a Walsh code that is used to have this channel to be distinguished from other traffic channels. It should be noted that the transmit modulator with capacity of M channels has M channel modulators, each with its own Walsh code. The Walsh codes used in the channel modulators are orthogonal to each other. For simplicity, only one traffic channel is depicted in the transmit modulator of FIG. 1.

The interleaved symbol output and the Walsh sequence are exclusive-OR'ed by Walsh code combiner 121. The chip rate of Walsh code becomes the CDMA spreading speed. The spread data stream is provided as input to a QPSK modulator comprising PN code combining unit 130, lowpass filter 140, analog signal modulator 150, and analog signal summer 160. The data stream from channel modulator 120 is inputted to each of PN code combiners 131 and 132 that multiply the data stream by in-phase and quadrature channel PN sequences, respectively. The two resulting data stream are provided as input to lowpass filter 140 for bandwidth reduction and are then modulated to analog signals through two mixers 151 and 152. The outputs of the mixers 151 and 152 are added into an analog signal by analog signal summer 160.

The output signal of analog signal summer 160 is modulated into an IF signal by IF signal modulator 170. Frequency up-converter 180 converts the IF signal to an RF signal and then the RF signal is amplified, bandpass filter, and radiated through an antenna system by RF transmitter 190.

When 64 Walsh codes are used for channelization, 64 channels are totally available in the transmit modulator of FIG. 1, because only one traffic channel is available for each channel modulator. Except the pilot, the sync, and the paging channels, 61 channels can be used for traffic channels. Therefore, about 30 channels can be at most be maintained in

the transmit modulator to support the good-quality communication service (It is known in the art that the number of channels of good-quality service is about 30 even though there are 61 available traffic channels). Hence, as the number of users increases, the channel resource that should be allocated to each user decreases. As a result, if the existing voice channels are used for data communication service, it is impossible to obtain more data channels, so that the data communication cost increases because users pay the same cost as that of the existing voice channels. It is reason why it is difficult to provide data communication service at a lower cost.

3. Disclosure of Invention

It is a primary object of the present invention to provide a method and apparatus for increasing the number of channels in a code division multiple access (CDMA) system by dividing a traffic channel into a plurality of subchannels of low data rate.

It is another object of the present invention to provide a method and apparatus for enabling reliable and long-distance communication with low power in a CDMA system by obtaining processing gain on the subchannels of low data rate in the demodulation process.

In a code division multiple access (CDMA) transmit modulator comprising a channel encoder for convolutionally encoding input signal from a vocoder with symbol repetition and interleaving the encoded signal; a channel modulator for combining the output signal from the channel encoder and an orthogonal code corresponding to a traffic channel; a pair of pseudo noise (PN) combiners, each for combining the modulated signal and a respective one of a pair of predetermined-offset PN signals; a pair of lowpass filters, each for filtering a respective output signal of the pair of

PN combiners and flattening the power level of the resulting signal; a pair of digital-to-analog converters, each for converting a respective output signal of the pair of lowpass filters to analog signal; and an analog signal summer for
5 summing the analog signals of the pair of digital-to-analog converters, an channel increasing apparatus according to the present invention comprises a plurality of subchannel encoders substituted for said channel encoder, each for convolutionally encoding with symbol repetition and
10 interleaving input data from a respective one of a plurality of subchannels, the data rate of each of the plurality of subchannels being lower than the encodable data rate of the traffic channel by said channel encoder; a plurality of subchannel modulators, each for combining an output signal
15 from a respective one of said plurality of subchannel encoders and a respective orthogonal code signal distinguishing one from another subchannel, all subchannels being accommodated in a single traffic channel; and a subchannel summer for summing output signals of said plurality of the subchannel
20 modulators and providing the summed signal to said channel modulator..

The apparatus according to the present invention provides the following advantages.

First, because each of traffic channels can be divided into
25 a plurality of subchannels of low data rate utilizing multiple modulation, the apparatus makes it possible to efficiently use the channel resource. Therefore, the apparatus enables to provide more data channels of low data rate and to reserve more voice channels for cellular phones as well.

30 Second, the increase of processing gain in the correlation detection by using multiple modulation of subchannels of low data rate enables data of low rate to be transmitted farther and more reliably with less power than in the transmit

modulator of the conventional art.

Third, in environments where data transmission using CDMA voice channels is nearly impossible, the apparatus enables data transmission of low rate and thus be utilized in very long distance communication and/or sea rescue communication systems.

4. Brief Description of Drawings

The accompanying drawings, which are included to provide a further understanding of the invention, illustrate the preferred embodiment of this invention, and together with the description, serve to explain the principles of the present invention.

In the drawings:

FIG. 1 is a block diagram of a CDMA transmit modulator;

FIG. 2 is a block diagram of a CDMA transmit modulator equipped with the apparatus according to an embodiment of the present invention; and

FIG. 3 is a block diagram of a subchannel summer shown in FIG. 2.

5. Modes for Carrying out the Invention

Hereinafter, a preferred embodiment of the present invention will be described in detail referring to the accompanying drawings.

FIG. 2 shows a block diagram of a transmit modulator including the apparatus according to an embodiment of the present invention. In the transmit modulator, the following constituting components: channel modulator 120, PN code combining unit 130, lowpass filter 140, analog signal modulator 150, analog signal summer 160, IF signal modulator 170, frequency up-converter 180, and RF transmitter 190 are all the same as those of the transmit modulator of FIG. 1. Hence, their numbers on FIG. 2 is the same as those of FIG. 1 and thus the description of their operations is omitted

here.

As shown in FIG. 2, the apparatus 200 according to an embodiment of the present invention comprises a plurality of subchannel encoders 210 (210-1, 210-2, ..., 210-N, where N is the number of subchannels), the function of each subchannel encoder being the same as that of channel encoder 110 of FIG. 1, a plurality of subchannel modulators 220 (220-1, 220-2, ..., 220-N), and a subchannel summer 230.

The input data on each subchannel is convolutionally encoded with repetition and then interleaved by a respective one of the plurality of subchannel encoders. It should be noted that each input data rate is lower than the encodable input data rate by channel encoder 110 of FIG. 1 (hereinafter, 19.2 kbps) divided by the number of subchannels. That is, the input data rate on each subchannel is lower than $19.2/N$ kbps. Each subchannel modulator combines data that is provided at a rate lower than $19.2/N$ kbps on the corresponding subchannel encoder and a respective orthogonal code of 19.2 kbps rate that is selected from an orthogonal code set, f_1 through f_n (code f_1 and code f_m are orthogonal each other if 1 is not equal to m). Hereinafter, such orthogonal codes are referred to as subchannel orthogonal sequences. Subchannel summer 230 sums the outputs of the plurality of subchannel modulators 220, each inputted at 19.2 kbps rate, and provides the summed output data to channel modulator 120.

Subchannel summer 230, as shown in FIG. 3, comprises a plurality of subchannel memories 231 (231-1, 231-2, ..., 231-N), each for storing the subchannel data from the corresponding subchannel modulator, and a data processor 232 for processing the data stored in the plurality of subchannel memories 231. It should be noted that subchannel summer 230 is capable of reducing the energy of the output of subchannel summer 230. To be specific, At a rate of 19.2 kbps, the output of each

subchannel modulator is inputted to subchannel summer 230. Summing each input data of 19.2 kbps increases the energy of the output of subchannel summer 230, which causes noise or cross-talk over other traffic channels. Therefore, it is
5 required to reduce the energy of each of the subchannel data.

Hereinafter, the operations of multiple modulation in the apparatus according to an embodiment of the present invention will be described in detail.

Input data on each subchannel is convolutionally encoded,
10 with symbol repetition, and interleaved by a respective one of the plurality of subchannel encoders and is then provided as input to a respective one of the plurality of subchannel modulators. It should be understood that it is required that the data rate of each subchannel is lower than $19.2/N$ kbps,
15 where N is the number of subchannels. For example, if subchannel orthogonal sequence used for each subchannel modulator is 64-bit long and is generated at a rate of 19.2 kbps, the data rate of each subchannel should be lower than 300 bps in order that data chip of 1 bit is multiplied by 64
20 bit-long subchannel orthogonal sequence of 19.2 kbps ($64/19.2$ kbps = $1/300$). This requirement of the data rate of each subchannel is satisfied when the number of subchannels is smaller than 64.

Each subchannel modulator multiplies the subchannel
25 encoded data by a respective one of subchannel orthogonal sequences, f_1, f_2, \dots, f_n , and provides the modulated subchannel data to subchannel summer 230. It should be noted that the bit rate of each subchannel orthogonal sequence is lower than the bit rate of Walsh codes, each being used in
30 channel modulator for distinguishing traffic channels in the frequency assignment (FA). For example, each subchannel orthogonal sequence of 19.2 kbps is multiplied with respective low speed input data to distinguish a subchannel

from others. All subchannels with subchannel orthogonal sequence multiplied are applied to subchannel summer 230. In this way, a plurality of subchannels of low data rate can be obtained within each traffic channel.

5 As shown in FIG. 3, the output of each subchannel modulator is stored in the corresponding subchannel memory 231 and is then provided as input to the subchannel data processor 232 in which the stored data are processed so as to reduce its energy level. In this data process, value 0 is considered as
10 -1. The output of the subchannel summer 230 requires more representation levels than two levels because the output should represent the sum of the outputs of the plurality of subchannels. For example, if the number of subchannels is 16, the output of the subchannel summer 230 should be represented
15 with 32 levels from -16 to +16. Hence, 5 bits are required to represent an output level of channel summer 230.

Channel modulator 120 combines several bits belonging to single output level of subchannel summer 230 and a Walsh code defining a traffic channel. Because the data rate of each
20 subchannel is M (M is an integer) times lower than that of the traffic channel to which the subchannels belong, the output of subchannel summer 320 is not varied over a period of the Walsh sequence. Therefore, the plurality of subchannels can be formed within each traffic channel. In
25 addition, a received signal gain is increased since data signal, whose energy is detected through a correlation process with a Walsh code, of a subchannel undergoes a correlation process again with a subchannel orthogonal sequence.

30 The output of channel modulator 120 in which each subchannel data is modulated is combined with each of a pair of PN codes by PN code combining unit 130. The pair of PN codes are generated with preset-offset phase such that multiple

cell-sites or sectors using the same frequency assignment are distinguished from one another. This PN code spreads several bits indicative of mixed value of all subchannel data.

Then, the outputs of PN code combining unit 130 are provided
5 as input to lowpass filter 140 that filters and flattens the outputs using a pair of digital FIR filters such that the outputs of lowpass filter 140 has flat power level in the frequency assignment. It should be noted that the plurality of subchannels belonging to a traffic channel are processed
10 simultaneously by the pair of digital FIR filters.

The digital signals of lowpass filter 140 are modulated into analog signals by analog signal modulator 150 that includes a pair of digital-to-analog converters and a pair of mixers, and are then provided to analog signal summer 160.
15 The analog signals are summed into an analog signal by analog signal summer 160 and the resulting analog signal is inputted to IF signal modulator 170. If it happens that the energy of the output of analog signal summer 150 is high due to the summation of signals of all traffic channels, the energy of
20 the output signal is reduced to an appropriate energy level by analog signal summer 150.

Receiving the output signal from analog signal summer 150, IF signal modulator 170 produces a modulated signal using QPSK modulation. In case of QPSK modulation, in-phase and
25 quadrature phase PN sequences are used in PN code combining unit 130, and sine and cosine functions are used in the pair of mixers of analog signal modulator 150.

The IF signal is up-converted into an RF signal by frequency up-converter 180. The RF signal is amplified by an RF
30 amplifier (not shown), and is then bandpass filtered and radiated through an antenna system (not shown).

The foregoing is provided only for the purpose of illustration and explanation of the preferred embodiments of

the present invention, so changes, variations and modifications may be made without departing from the spirit and scope of the invention.

CLAIMS

1. In a code division multiple access transmit modulator comprising a channel encoder for convolutionally encoding input signal from a vocoder with symbol repetition and
5 interleaving the encoded signal; a channel modulator for combining the output signal from said channel encoder and an orthogonal code signal distinguishing one from another traffic channel; a pair of pseudo noise (PN) combiners, each for combining the output signal of said channel modulator and
10 a respective one of a pair of pseudo noise signals which have a predetermined offset in phase; a pair of lowpass filters, each for filtering a respective output signal of said plurality of PN combiners and flattening the power level of the output signal; and an analog signal modulator for
15 converting the output signals of said pair of lowpass filters to an RF signal, an apparatus for obtaining multiple subchannels within a traffic channel, comprising:

a plurality of subchannel encoders substituted for said channel encoder, each for convolutionally encoding with
20 symbol repetition and interleaving input data from a respective one of a plurality of subchannels, the data rate of each of the plurality of subchannels being lower than the encodable data rate of the traffic channel by said channel encoder;

25 a plurality of subchannel modulators, each for combining an output signal from a respective one of said plurality of subchannel encoders and a respective orthogonal code signal distinguishing one from another subchannel, all subchannels being accommodated in a single traffic channel; and

30 a subchannel summer for summing output signals of said

plurality of the subchannel modulators and providing the summed signal to said channel modulator.

2. An apparatus according to claim 1, wherein the data rate of each of the plurality of subchannels is N times lower than a predetermined data rate of input signal that is inputted to said channel encoder, N being the number of said subchannel encoders.

3. An apparatus according to claim 1, wherein the data rate of said orthogonal code signal defining a subchannel is equal to a predetermined data rate of input signal that is inputted to said channel modulator.

4. An apparatus according to claim 1, wherein said subchannel summer comprises:

a plurality of storing means, each for storing subchannel signal from a respective one of said plurality of subchannel modulators; and

data processing means for reading and processing the subchannel signals stored in said plurality of storing means.

5. An apparatus according to claim 1 or claim 4, wherein said subchannel summer reduces the energy of the subchannel data of each or all of the plurality of subchannels.

6. A method of obtaining multiple channels within a traffic channel in a code division multiple access transmit modulator, comprising the steps of:

(a) encoding a plurality of input signals by using convolutional encoding, symbol repetition, and interleaving independently;

(b) multiplying each of a plurality of the encoded signals by a first respective orthogonal code signal distinguishing one from another subchannel, so as to provide a plurality of resultant subchannelized input signals;

(b) mixing the plurality of subchannelized input signals into a resultant combined signal;

(d) multiplying the combined signal by a second orthogonal code signal distinguishing one from another traffic channel, so as to provide a resultant channelized signal;

(e) multiplying the channelized signal by a PN code which is predetermined-offset in phase, so as to provide a PN code modulated signal;

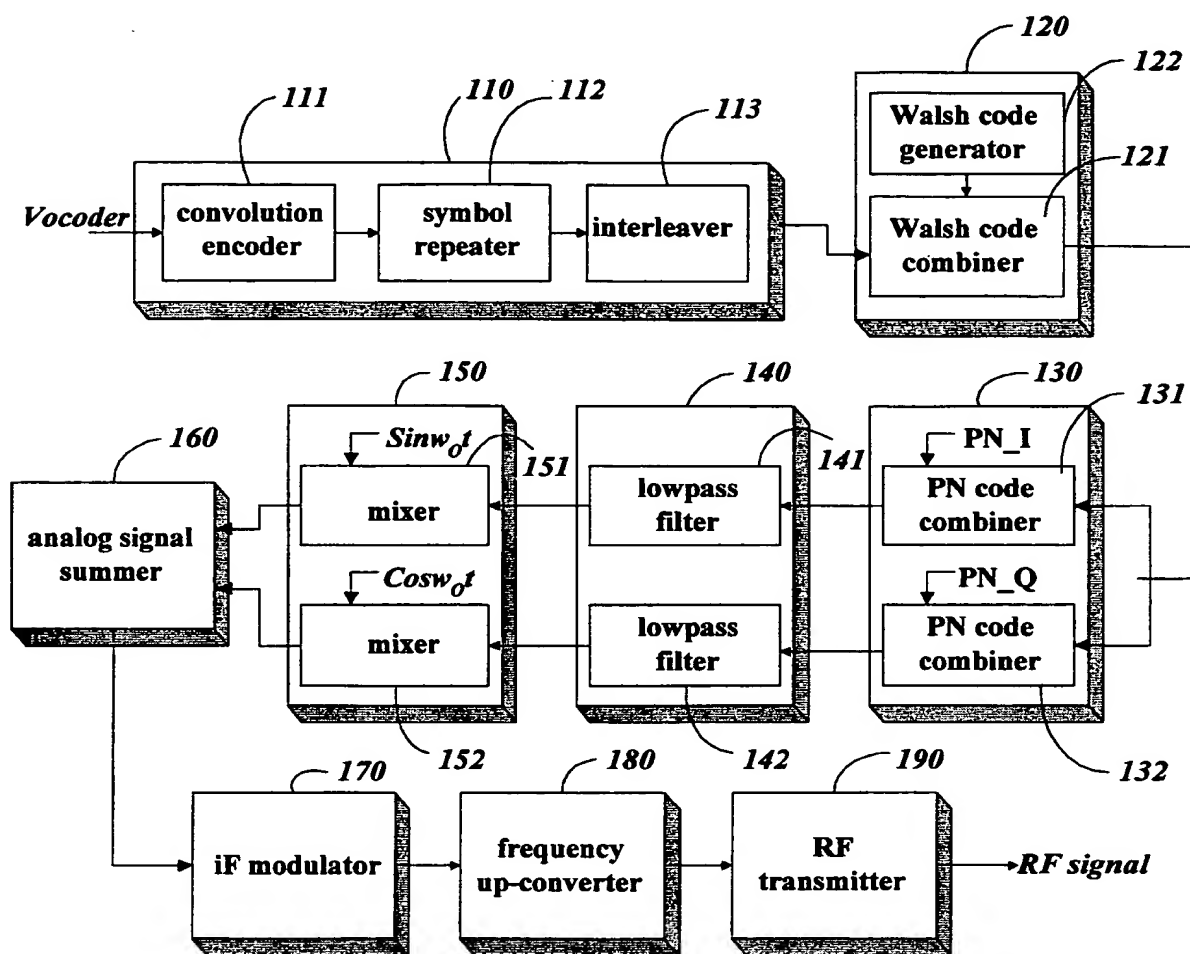
(f) filtering the PN code modulated signal and flattening the power level in the frequency band; and

(g) converting the filtered signal into an radio frequency signal.

7. A method according to claim 6, wherein the data rate of the input signal is N times lower than the data rate defined for the resultant combined signal, N being the number of said plurality of input data.

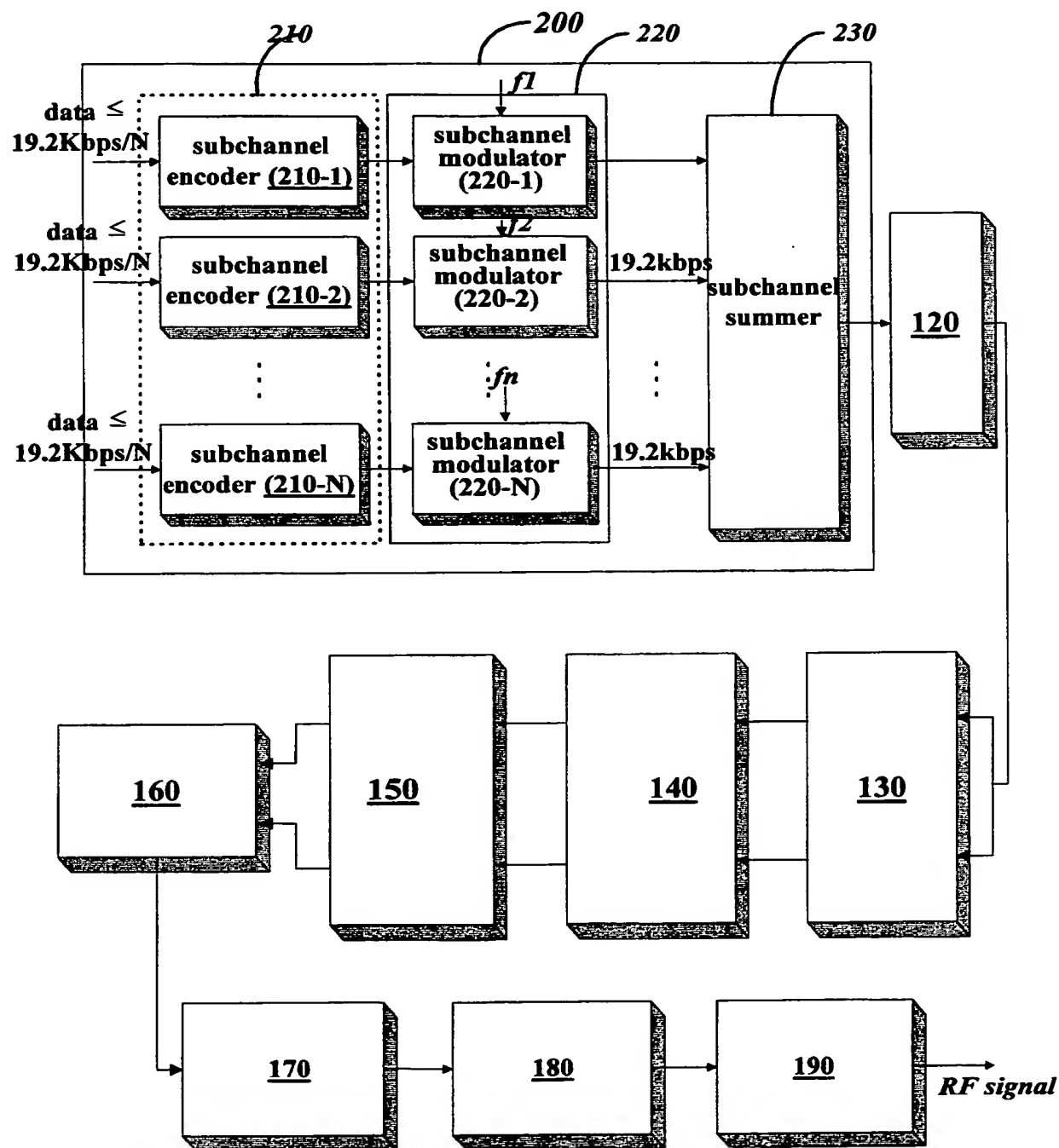
8. A method according to claim 7, wherein the bit rate of the first orthogonal code signal is equal to the data rate defined for the resultant combined signal.

FIG. 1



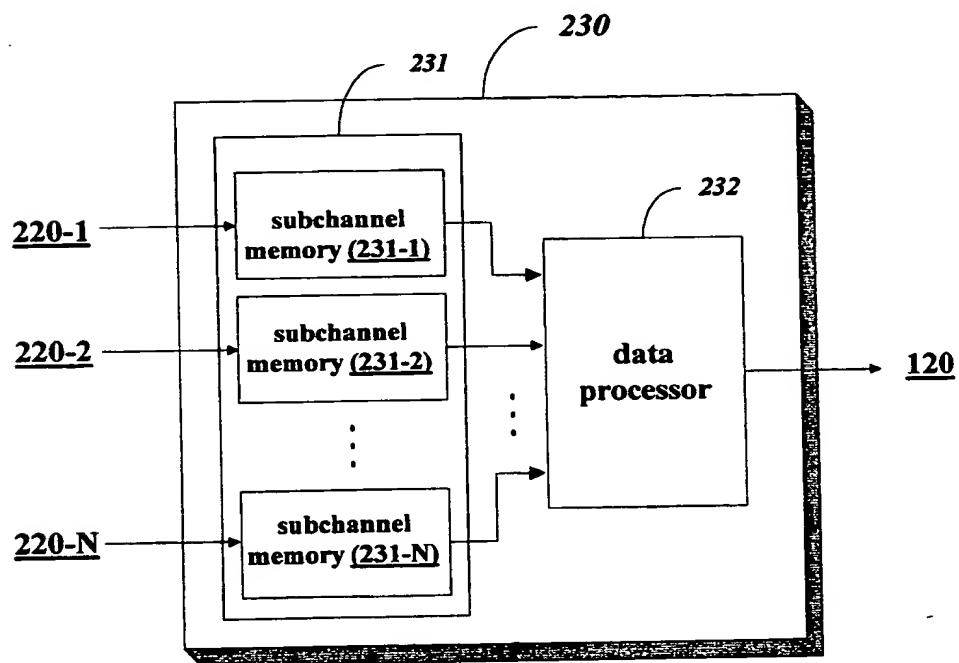
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FIG. 2



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FIG. 3



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INTERNATIONAL SEARCH REPORT

International application No.

PCT/KR00/01088

A. CLASSIFICATION OF SUBJECT MATTER**IPC7 H04B 1/69**

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC7 H04B 1/69, 1/707, 7/26

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practicable, search terms used)
search terms : CDMA system, multichannel, multicode, subchannel, orthogonal code**C. DOCUMENTS CONSIDERED TO BE RELEVANT**

Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	US 5442625 A 15 Aug. 1995 (AT&T) column 1 lines 43 to column 1 lines 59	1, 2, 3, 6, 7, 8
Y	US 5586113 A 17 Dec. 1996 (NTT Mobile Communication Network Inc.) column 2 lines 56 to column 5 lines 59 (1, 6)	1, 6
Y	US 5859840 A 12 Jan. 1999 (Qualcomm Inc.) column 2 lines 26 to column 3 lines 24 (1, 6)	1, 6
Y	US 5914950 A 22 June 1999 (Qualcomm Inc.) column 4 lines 37 to column 5 lines 27 (1, 6)	1, 6
Y	K. Okawa, F. Adachi, "Orthogonal multi-spreading factor forward link for coherent DS-CDMA mobile radio", IEEE 6th International Conference on Universal Personal Communications Record vol.2, pp618 - 622, 12-16 Oct. 1997 (1, 2, 3, 6, 7, 8)	1, 2, 3, 6, 7, 8

☐ Further documents are listed in the continuation of Box C.☒ See patent family annex.

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INTERNATIONAL SEARCH REPORT

Information on patent family members

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Patent document cited in search report	Publication date	Patent family member(s)	Publication date
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